

Outline

1. Introduction of Frequency Response Concept
 2. Frequency Response of the Typical Elements
 3. Bode Diagram of Open-loop System
 4. Nyquist-criterion
 5. Frequency Response Based System Analysis
 6. Frequency Response of Closed-loop Systems
- Graphical method
- Analysis
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6-1 Introduction

Key problem of control: stability and system performance

Question: Why Frequency Response

Analysis:

- 1. Weakness of root locus method relies on the existence of open-loop transfer function**
 - 2. Weakness of time-domain analysis method is that time response is very difficult to obtain**
 - Computational complex**
 - Difficult for higher order system**
 - Difficult to partition into main parts**
 - Not easy to show the effects by graphical method**
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Frequency Response Analysis

Three advantages:

- * Frequency response(mathematical modeling) can be obtained directly by experimental approaches.
- * Easy to analyze effects of the system with sinusoidal signals
- * Convenient to measure system sensitivity to noise and parameter variations

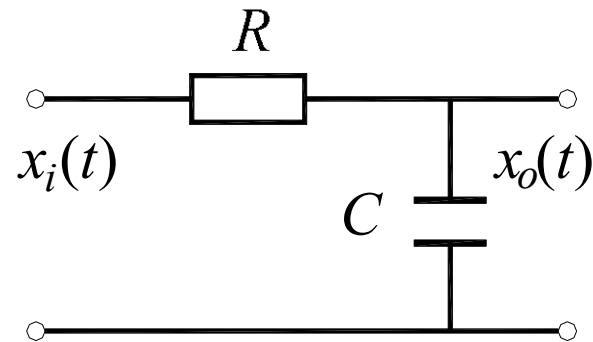
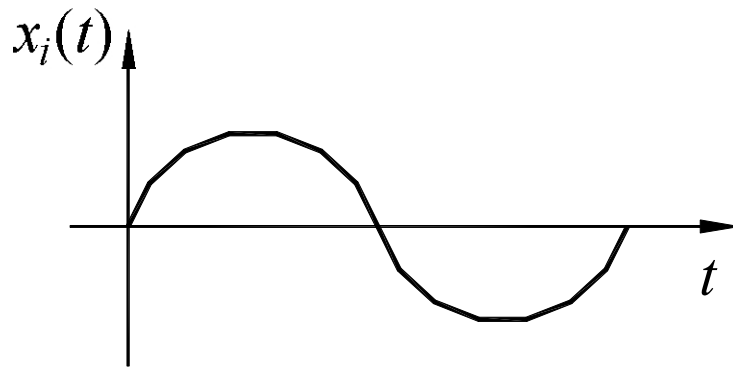
However, NEVER be limited to sinusoidal input

* **Frequency-domain performances → time-domain performances**

Frequency domain analysis is a kind of (indirect method) engineering method. It studies the system based on frequency response which is also a kind of mathematical model.

Frequency Response

Example 5.1: RC circuit



$$u_i = A \sin(\omega t + \phi_0)$$

$$U_i(s) = \frac{U_{1m}}{s^2 + \omega^2}$$

Transfer function

$$U_o(s) = \frac{1}{Ts + 1} \cdot \frac{U_{1m}}{s^2 + \omega^2}$$

$$G(s) = \frac{1}{RCs + 1} = \frac{1}{Ts + 1}$$

By using inverse Laplace transform

$$u_0(t) = \underbrace{\frac{U_{1m} T \omega}{1 + T^2 \omega^2} e^{-\frac{t}{T}}}_{\text{Transient response}} + \underbrace{\frac{U_{1m}}{\sqrt{1 + T^2 \omega^2}} \sin(\omega t + \varphi)}_{\text{Steady state response}} \quad \varphi = -\arctg(\omega T)$$

Steady state response of u_0

$$\begin{aligned} \lim_{t \rightarrow \infty} u_0 &= \frac{U_{1m}}{\sqrt{1 + T^2 \omega^2}} \sin(\omega t + \varphi) \\ &= U_{1m} \left| \frac{1}{1 + j\omega T} \right| \sin(\omega t + \angle \frac{1}{1 + j\omega T}) \end{aligned}$$

Proposition: When the input to a linear time-invariant (LTI) system is sinusoidal, the steady-state output is a sinusoid with the same frequency but possibly with different amplitude and phase.

Definition: *Frequency response (or characteristic)* is the ratio of the complex vector of the steady-state output versus sinusoidal input for a linear system.

$$G(j\omega) = \frac{1}{1 + j\omega T} = A(\omega) e^{j\phi(\omega)}$$

$$A(\omega) = \left| \frac{1}{1 + j\omega T} \right| = \frac{1}{\sqrt{1 + \omega^2 T^2}};$$

Magnitude response

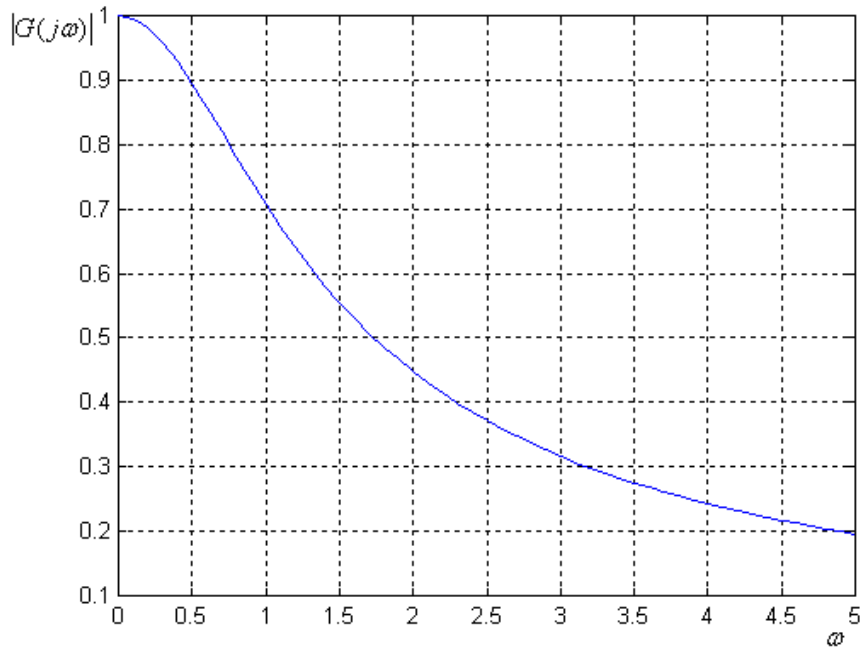
$$\phi(\omega) = \angle \left(\frac{1}{1 + j\omega T} \right) = -\arctg \omega T$$

Phase response

$\omega = 0$ The output has same magnitude and phase with input

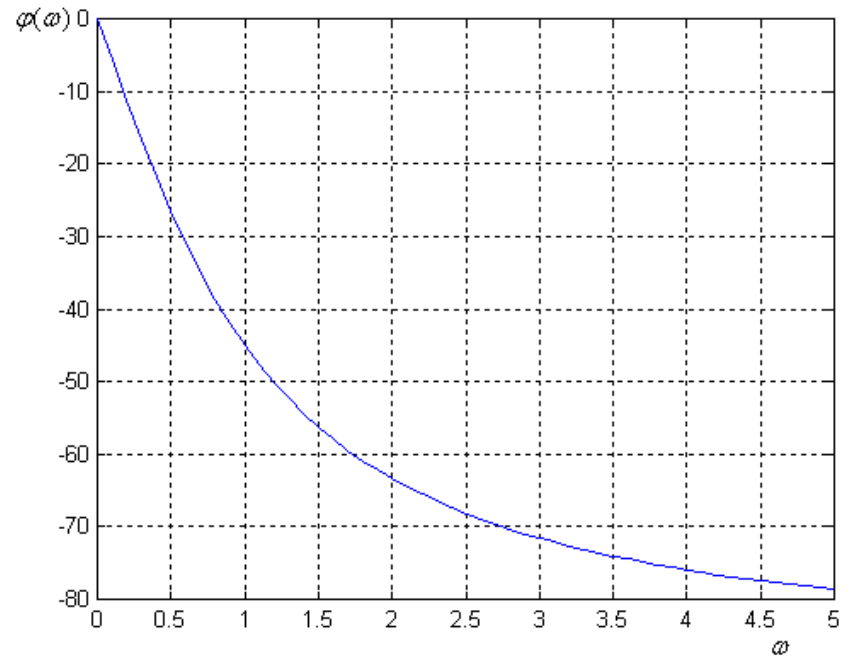
Magnitude will be attenuated and phase lag is increased.

ω



Magnitude of output versus input

Magnitude characteristic



Phase error of output and input

Phase characteristic



Generalized to linear time-invariant system

Transfer function of closed-loop system

$$G(s) = \frac{C(s)}{R(s)} = \frac{N(s)}{D(s)} = \frac{N(s)}{(s + p_1)(s + p_2)\dots(s + p_n)}$$

where p_1, \dots, p_n are different closed-loop poles.

Given the sinusoidal input

$$r(t) = A_r \sin \omega t \quad R(s) = \frac{A_r \omega}{s^2 + \omega^2}$$

$$\begin{aligned} C(s) &= G(s) \cdot \frac{A_r \omega}{s^2 + \omega^2} = \frac{N(s)}{D(s)} \cdot \frac{A_r \omega}{s^2 + \omega^2} \\ &= \frac{a}{s + j\omega} + \frac{\bar{a}}{s - j\omega} + \frac{b_1}{s + p_1} + \frac{b_2}{s + p_2} + \dots + \frac{b_n}{s + p_n} \end{aligned}$$

$$c(t) = ae^{-j!t} + \bar{a}e^{j!t} + b_1e^{-p_1t} + b_2e^{-p_2t} + \dots + b_n e^{-p_nt}$$

$$= \sum_{i=1}^n b_i e^{-p_i t} + (ae^{-j!t} + \bar{a}e^{j!t})$$

$$= c_t(t) + c_s(t)$$

(t ! 0)

Transient response Steady state response

For a stable closed-loop system, we have $-p_i < 0$

$$a = G(s) \# \frac{A_r!}{s^2 + !^2} \# (s + j!) \Big|_{s=-j!} = -\frac{A_r G(-j!)}{2j}$$

$$\bar{a} = G(s) \cdot \frac{A_r!}{s^2 + !^2} \cdot (s - j!) \Big|_{s=j!} = \frac{A_r G(j!)}{2j}$$

$$G(j!) = |G(j!)| e^{j\angle G(j!)} \quad G(-j!) = |G(-j!)| e^{-j\angle G(j!)} = |G(j!)| e^{-j\angle G(j!)}$$

Furthermore, we have

$$\begin{aligned}c_s(t) &= ae^{j\omega t} + \bar{a}e^{-j\omega t} \\&= A_r |G(j\omega)| \frac{e^{j(\omega t + \angle G(j\omega))} + e^{-j(\omega t + \angle G(j\omega))}}{2} \\&= A_r |G(j\omega)| \sin(\omega t + \angle G(j\omega)) \\&= A_c \sin(\omega t + \varphi)\end{aligned}$$

The magnitude and phase of steady state are as follows

$$A_c = A_r |G(j\omega)|; \quad \varphi = \angle G(j\omega)$$

By knowing the transfer function $G(s)$ of a linear system, the magnitude and phase characteristics completely describe the steady-state performance when the input is sinusoid.

Frequency-domain analysis can be used to predict both time-domain transient and steady-state system performance.

Relation of transfer function and frequency characteristic of LTI system (only for LTI system)

$$G(s) \Big|_{s=j\omega} = G(j\omega) = |G(j\omega)| e^{j\angle G(j\omega)}$$

➤ Substitute $s=j\omega$ into the transfer function

$$F(s) = L[f(t)] = \int_{-\infty}^{+\infty} f(t) e^{st} dt$$

Laplace transform

$$f(t) = L^{-1}[F(s)] = \int_{c-j\infty}^{c+j\infty} F(s) e^{st} ds$$

Inverse Laplace transform

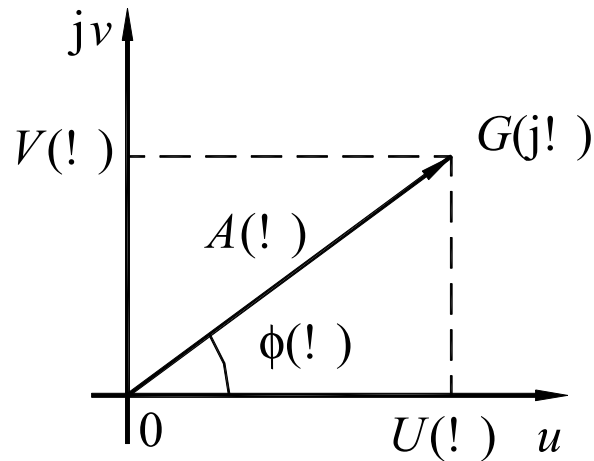
$$F(j\omega) = F[f(t)] = \int_{-\infty}^{+\infty} f(t) e^{-j\omega t} dt$$

Fourier transform

$$f(t) = F^{-1}[F(s)] = \frac{1}{2\pi j} \int_{-\infty}^{+\infty} F(j\omega) e^{j\omega t} d\omega$$

Inverse Fourier transform

Vector of frequency characteristics



It is a complex vector, and has three forms:

Algebraic form

$$G(j\omega) = U(\omega) + jV(\omega)$$

Polar form

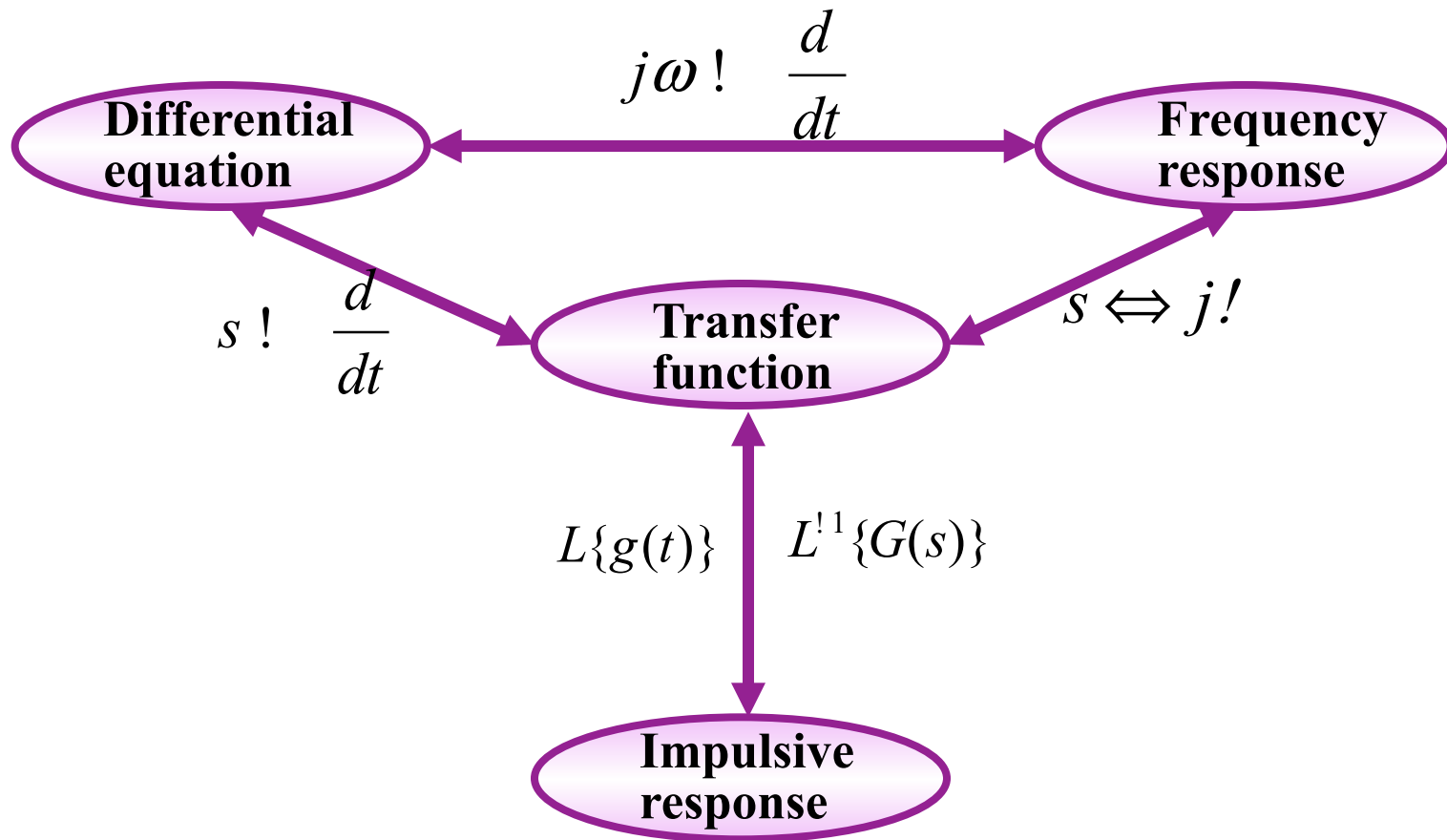
$$G(j\omega) = |G(j\omega)| \angle \phi(\omega) = A(\omega) \angle \phi(\omega)$$

Exponential form $G(j\omega) = |G(j\omega)| e^{j\phi(\omega)} = A(\omega) e^{j\phi(\omega)}$

$$A(\omega) = |G(j\omega)| = \sqrt{U^2(\omega) + V^2(\omega)} \quad U(\omega) = A(\omega) \cos \phi(\omega)$$

$$\phi(\omega) = \arctan \frac{V(\omega)}{U(\omega)} \quad V(\omega) = A(\omega) \sin \phi(\omega)$$

We have learned following mathematical models:
differential equation, transfer function and frequency response



Example 5.2: Given the transfer function

$$G(s) = \frac{C(s)}{R(s)} = \frac{1}{s^2 + 3s + 4}$$

Differential equation:

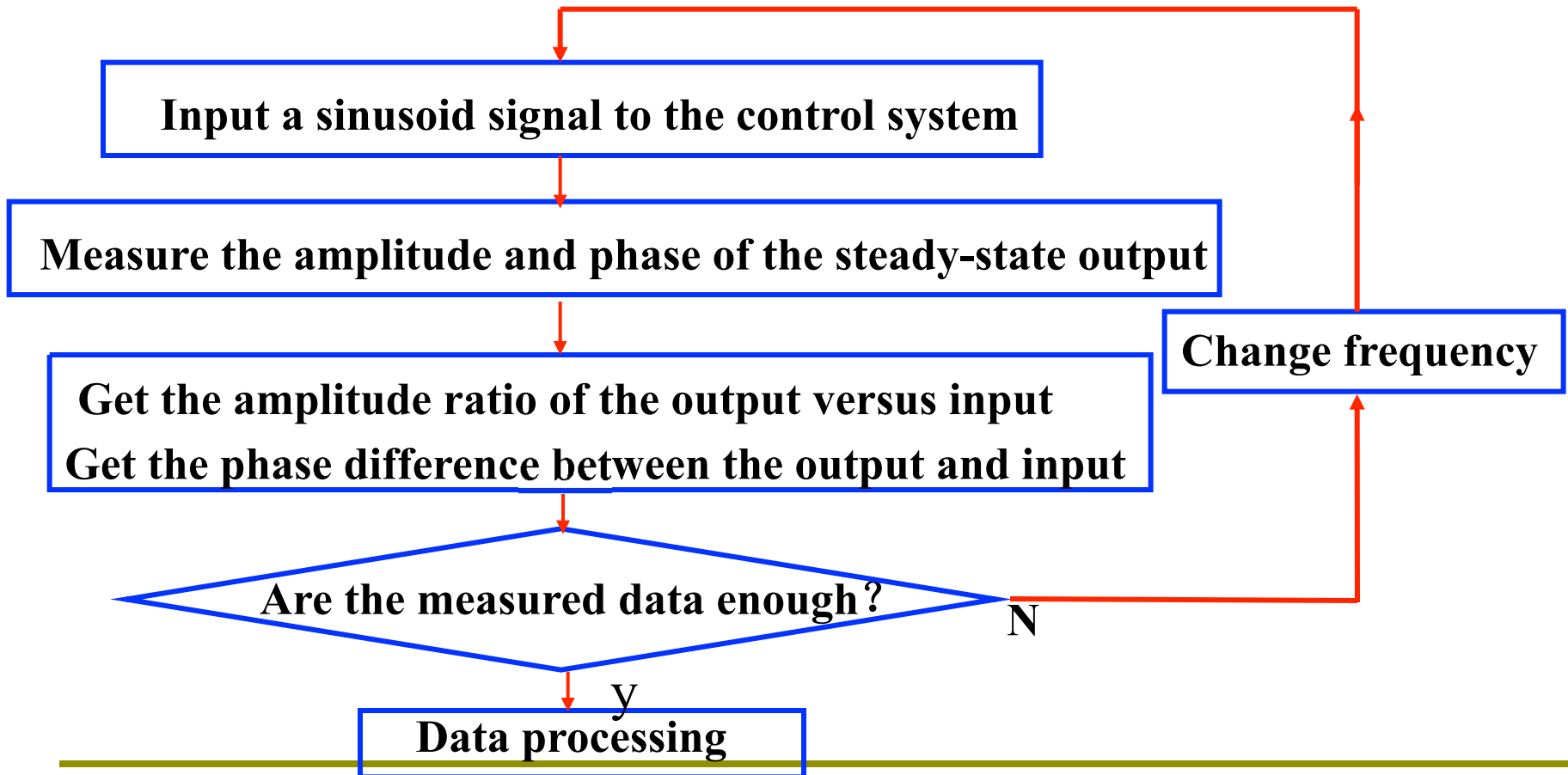
$$\frac{d^2c(t)}{dt^2} + 3\frac{dc(t)}{dt} + 4c(t) = r(t)$$

Frequency response:

$$G(j\omega) = \frac{c(j\omega)}{s(j\omega)} = \frac{1}{(j\omega)^2 + 3(j\omega) + 4} = \frac{1}{4 - \omega^2 + 3j\omega}$$

6-2 Frequency Characteristics of Typical Elements of system

How to get frequency characteristic?



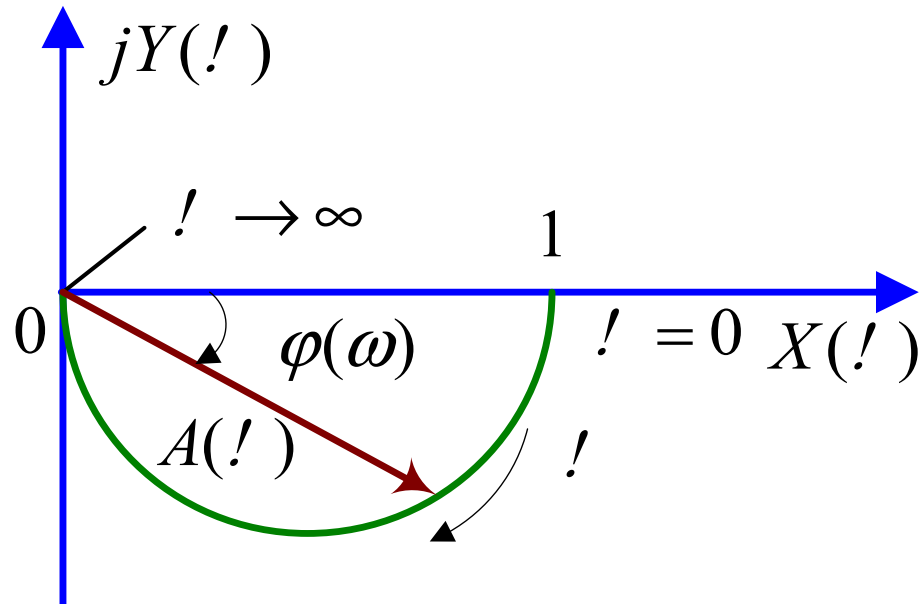
Nyquist Diagram

- Magnitude characteristic diagram
A-! plot
 - Phase characteristic diagram
X-! plot
 - Gain-phase characteristic diagram — Nyquist diagram
- } Bode diagram

Polar form or algebraic form: *A* and *X* define a vector for a particular frequency ! .

$$G(j\omega) = |G(j\omega)| e^{j\varphi(\omega)} = X(\omega) + jY(\omega)$$

ω	0	$1/(2RC)$	$1/RC$	$2/RC$	$3/RC$	$4/RC$	$5/RC$	∞
$A(\omega)$	1	0.89	0.707	0.45	0.32	0.24	0.2	0
$\phi(\omega)$	0	-26.6	-45	-63.5	-71.5	-76	-78.7	-90



Nyquist Diagram of RC Circuit

Bode Diagram

- **Bode Diagram: Logarithmic plots of magnitude response and phase response**
- **Horizontal axis: $\lg!$ (logarithmic scale to the base of 10) (unit: rad/s)**
- **Log Magnitude**

In feedback-system, the unit commonly used for the logarithm of the magnitude is the **decibel** (dB)

$$L(\omega) = 20 \lg |G(j\omega)| = 20 \lg A(\omega)$$

Property 1: As the frequency doubles, the decibel value increases by 6 dB.

As the frequency increases by a factor of 10, the decibel value increases by 20 dB.

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